Core loss is generated by the changing magnetic flux field within a material, since no magnetic materials exhibit perfectly efficient magnetic response. Core loss density (PL) is a function of half of the AC flux swing ( $\frac{1}{2}\Delta B=B_{pk}$ ) and frequency (f). It can be approximated from core loss charts or the curve fit loss equation:

$$PL = aB_{pk}^{b} f^{c}$$

where a, b, c are constants determined from curve fitting, and  $B_{pk}$  is defined as half of the AC flux swing:

$$B_{\text{pk}} = \frac{\triangle B}{2} = \frac{B_{\text{AC max}} - B_{\text{AC min}}}{2}$$

Units typically used are (mW/cm $^3$ ) for PL; Tesla (T) for B<sub>pk</sub>; and (kHz) for f.

The task of core loss calculation is to determine  $B_{pk}$  from known design parameters.

# Method 1 – Determine $B_{pk}$ from DC Magnetization Curve. $B_{pk} = f(H)$

Flux density (B) is a non-linear function of magnetizing field (H), which in turn is a function of winding number of turns (N), current (I), and magnetic path length ( $I_{\rm p}$ ). The value of B<sub>pk</sub> can typically be determined by first calculating H at each AC extreme:

$$\begin{split} H_{\text{ACmax}} &= \left[ \frac{N}{I_{\text{e}}} \bigg( \ I_{\text{DC}} + \frac{\Delta I}{2} \bigg) \right] \\ H_{\text{ACmin}} &= \left[ \frac{N}{I_{\text{e}}} \bigg( \ I_{\text{DC}} - \frac{\Delta I}{2} \bigg) \right] \end{split}$$

Units typically used are (A·T/cm) for H. From H<sub>AC max</sub>, H<sub>AC min</sub>, and the BH curve or equation (listed as DC Magnetization, pgs. 47 - 50)  $\rm B_{AC \, max}$ ,  $\rm B_{AC \, min}$  and therefore B<sub>nk</sub> can be determined.

#### 60μ Kool Mμ DC Magnetization (Example 2)



### Example 1 - AC current is 10% of DC current:

Approximate the core loss of an inductor with 20 turns wound on Kool M $\mu$  p/n 77894A7 p. 76 (60 $\mu$ ,  $I_e$ =6.35 cm,  $A_e$ =0.654 cm²,  $A_L$ =75 nH/T²). Inductor current is 20 Amps DC with ripple of 2 Amps peak-peak at 100kHz.

1.) Calculate H and determine B from BH curve (p. 48) or curve fit equation (p. 50):

$$\begin{array}{lll} H_{\text{ACmax}} &=& \frac{20}{6.35} \left(20 + \frac{2}{2}\right) = & 66.14 & \text{A-T}_{\text{cm}} & \rightarrow & B_{\text{ACmax}} \cong 0.40T \\ H_{\text{ACmin}} &=& \frac{20}{6.35} \left(20 - \frac{2}{2}\right) = & 59.84 & \text{A-T}_{\text{cm}} & \rightarrow & B_{\text{ACmin}} \cong 0.37T \end{array} \\ \rightarrow B_{\text{pk}} = \frac{\triangle B}{2} = \frac{0.40 - 0.37}{2} = 0.015T$$

2.) Determine Core Loss density from chart or calculate from loss equation p. 46:

$$PL = (62.65)(0.015^{1.781})(100^{1.36}) \approx 18.5 \frac{mW}{cm^3}$$

3.) Calculate core loss:

$$P_{fe} = (PL) (I_e) (A_e) \sim (18.5) (6.35) (0.654) \cong 77 \text{mW}$$

MAGNETICS

#### Example 2 - AC current is 40% of DC current:

Approximate the core loss for the same 20-turn inductor, with same inductor current of 20 Amps DC but ripple of 8 Amps peak-peak at 100kHz.

1.) Calculate H and determine B from BH curve fit equation p. 50:

$$\begin{array}{lll} H_{\text{ACmax}} = & \frac{20}{6.35} \left(20 + \frac{8}{2}\right) = & 75.59 & \text{A-T/cm} & \rightarrow & B_{\text{ACmax}} \cong 0.44T \\ H_{\text{ACmin}} = & \frac{20}{6.35} \left(20 - \frac{8}{2}\right) = & 50.39 & \text{A-T/cm} & \rightarrow & B_{\text{ACmin}} \cong 0.33T \end{array} \\ \rightarrow B_{\text{pk}} = \frac{\triangle B}{2} = \frac{0.44 - 0.33}{2} = 0.055T$$

- 2.) Determine Core Loss density from chart or calculate from loss equation p. 46:  $PL = (62.65)(0.055^{1.781})(100^{1.36}) \approx 188 \frac{mW}{cm^3}$
- 3.) Calculate core loss:  $P_{fe} = (PL) (I_e) (A_e) = (188) (6.35) (0.654) \approx 781 \text{mW}$

Note: Core losses result only from AC excitation. DC bias applied to any core does not cause any core losses, regardless of the magnitude of the bias.

#### Example 3 - pure AC, no DC:

Approximate the core loss for the same 20-turn inductor, now with 0 Amps DC and 8 Amps peak-peak at 100kHz.

1.) Calculate H and determine B from BH curve fit equation p. 50:

$$\begin{array}{lll} H_{\text{ACmax}} = & \frac{20}{6.35} \left( + \frac{8}{2} \right) = & 12.60 \ ^{\text{AT}} / \text{cm} & \rightarrow & \text{B}_{\text{ACmax}} \cong 0.092\text{T} \\ H_{\text{ACmin}} = & \frac{20}{6.35} \left( - \frac{8}{2} \right) = & -12.60 \ ^{\text{AT}} / \text{cm} & \rightarrow & \text{B}_{\text{ACmin}} \cong -0.092\text{T} \end{array}$$

Note: Curve fit equations are not valid for negative values of B. Evaluate for the absolute value of B, then reverse the sign of the resulting H value.

2.) Determine Core Loss density from chart or calculate from loss equation p. 46.

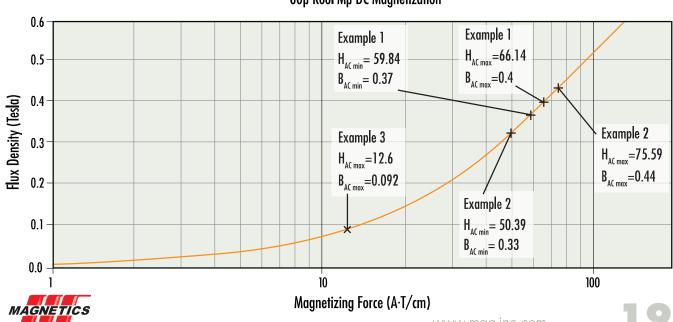
$$PL = (62.65)(0.092^{1.781})(100^{1.36}) \approx 470 \frac{mW}{cm^3}$$

3.) Calculate core loss:  $P_{fe} = (PL) (I_e) (A_e) = (470) (6.35) (0.654) \approx 1.95W$ 

Plotted below are the operating ranges for each of the three examples.

Note the significant influence of DC bias on core loss, comparing Example 3 with Example 2. Lower permeability results in less B<sub>pk</sub>, even if the current ripple is the same. This effect can be achieved with DC bias, or by selecting a lower permeability material.





Method 2, for small  $\triangle H$ , approximate  $B_{pk}$  from effective perm with DC bias.  $B_{pk} = f(\mu_e, \triangle H)$ 

The instantaneous slope of the BH curve is defined as the absolute permeability, which is the product of permeability of free space  $(\mu_0=4\pi \times 10^{-7})$  and the material permeability  $(\mu)$ , which varies along the BH curve. For small AC, this slope can be modeled as a constant throughout AC excitation, with  $\mu$  approximated as the effective perm at DC bias  $(\mu_0)$ :

$$\frac{dB}{dH} = \mu_0 \mu_e \quad \rightarrow \frac{\triangle B}{\triangle H} = \mu_0 \mu_e \quad \rightarrow \quad \triangle B = \mu_0 \mu_e \triangle H \quad B_{pk} = \frac{\triangle B}{2} = (0.5) \, \mu_0 \mu_e \triangle H$$

The effective perm with DC bias is shown in this catalog as % of initial perm and can be obtained from the DC bias curve or curve fit equation, pgs 29 - 34

$$B_{pk} = (0.5)(\mu_0)(\%\mu_i)(\mu_i)(100)(\triangle H)$$
 where  $\triangle H = \frac{N\triangle I}{I_0}$ 

 $\triangle H$  is multiplied by 100 because  $I_e$  is expressed in cm, while  $B_{pk}$  units include m.

Reworking Example 1 (20 Amps DC, 2 Amps pk-pk)

$$H_{DC} = \left[\frac{20}{6.35}(20)\right] = 63^{A.T}$$
 from curve or curve fit equation,  $\%\mu_1 = 0.58$ 

$$\mu_{i} = 60$$

$$\triangle H = \frac{N \triangle I}{I_{e}} = \frac{20(2)}{6.35} = 6.3^{AT} / \text{cm}$$

 $B_{pk} = 0.5(4\pi \times 10^{-7})(0.58)(60)(100)(6.3) \cong 0.014T$  (this compares to 0.015T using Method 1)

Reworking Example 2 (20 Amps DC, 8 Amps pk-pk)

From example 1,

$$H_{DC} = 63 \text{ A-T/cm}, \% \mu_i = 0.58; \mu_i = 60$$

$$\triangle H = \frac{N\triangle I}{\mathit{I_e}} = \frac{20(8)}{6.35} = 25.2 \, \text{A-T/cm}$$

 $B_{\text{pk}} = 0.5 \, (4\pi \times 10^{-7}) \, (0.58) \, (60) \, (100) \, (25.2) = 0.055 T \qquad \text{(this compares to 0.055T using Method 1)}$ 

Reworking Example 3 (0 Amps DC, 8 Amps pk-pk)

From example 2,

$$\triangle H = 25.20^{\text{A} \cdot \text{T/m}}$$

$$H_{DC} = 0^{A \cdot T} / _{cm}$$
 % $\mu_i = 1$ 

 $B_{pk} = 0.5(4\pi \times 10^{-7})(1)(60)(100)(25.2) = 0.095T$  (this compares to 0.092T using Method 1)



Method 3, for small  $\triangle H$ , determine  $B_{pk}$  from biased inductance.  $B_{pk=}=f(L,I)$ 

B can be rewritten in terms of inductance by considering Faraday's equation and its effect on inductor current:

$$V_L = NA \frac{dB}{dt} = L \frac{dI}{dt} \rightarrow dB = \frac{L}{NA} dI$$

L varies non-linearly with I. For small AC, L can be assumed constant throughout AC excitation and is approximated by the biased inductance (LDC).

$$\triangle B = \frac{L_{\text{DC}} \triangle I}{NA} \quad \rightarrow \quad B_{\text{pk}} = \frac{L_{\text{DC}} \triangle I}{2NA_{\text{e}}}$$

Another way of looking at this is by rewriting the relationship between B and L as:

$$\rightarrow \frac{dB}{dH} = \frac{L}{NA} \frac{dI}{dH}$$

Substituting (dH/dI) with (N/ $I_{e}$ ) and A with A $_{e}$ :

$$\rightarrow \frac{dB}{dH} = \frac{L I_e}{N^2 A_e}$$

L varies non-linearly with H. For small AC, the slope of the BH curve is assumed constant throughout AC excitation, and L is approximated by the biased inductance ( $L_{DC}$ ).

$$\frac{\triangle B}{\triangle H} = \frac{L_{\text{DC}}I_{\text{e}}}{N^{2}A_{\text{e}}} \quad \rightarrow \quad \triangle B = \frac{L_{\text{DC}}I_{\text{e}}}{N^{2}A_{\text{e}}} \ \triangle H = \frac{L_{\text{DC}}\triangle I}{NA_{\text{e}}} \quad \rightarrow \quad \triangle B_{\text{pk}} = \frac{L_{\text{DC}}\triangle I}{2NA_{\text{e}}}$$



Reworking Example 1:

$$\begin{split} L_{\text{rl}} \left( \text{no load} \right) &= \left( A_{\text{L}} \right) \left( N^2 \right) = \left( 75 \, \text{nH/T}^2 \right) \left( 20^2 \right) = 30 \mu \text{H} \\ L_{\text{DC}} \left( 20 A \right) &= \left( \% \mu_{\text{l}} \right) \left( L_{\text{rl}} \right) = \left( 0.58 \right) \left( 30 \right) = 17.4 \mu \text{H} \\ \rightarrow B_{\text{pk}} &= \frac{\left( 17.4 \right) \left( 10^{-6} \right) \left( 2 \right)}{2 \left( 20 \right) \left( 0.654 \right) \left( 10^{-4} \right)} = 0.013 T \qquad \text{(this compares to 0.015T per Method 1, 0.014T per Method 2)} \,. \end{split}$$

Reworking Example 2:

$$From \ example \ 1, \ L_{DC} = 17.4 \mu H$$
 
$$\rightarrow \ B_{pk} = \frac{(17.4) \, (10^{-6}) \, (8)}{2 \, (20) \, (0.654) \, (10^{-4})} = 0.053T \qquad \text{(this compares to } 0.055T \ per \ Method \ 1, 0.055T \ per \ Method \ 2).}$$

Reworking Example 3:

$$L_{\text{DC}} = L_{\text{nl}} = 30 \mu H$$
 
$$\rightarrow B_{\text{pk}} = \frac{(30) \, (10^{-6}) \, (8)}{2 \, (20) \, (0.654) \, (10^{-4})} = 0.092 T \qquad \text{(this compares to 0.092T per Method 1, 0.095T per Method 2)}.$$

The plot below illustrates the difference between Method 1 and Method 2

#### 60µ Kool Mµ DC Magnetization 0.47 0.45 Method 2 Method 2 Method 2 0.43 Flox Density (lesla) 0.39 0.37 0.35 0.33 Method 1 Method 1 $\mathsf{H}_{\mathsf{AC\ max}}$ H<sub>AC min</sub> 0.31 0.29 50 55 70 75 45 80 Magnetizing Force (A·T/cm)